



COMPLETE TV SOUND CHANNEL

The TDA3190 is a monolithic integrated circuit in a 16-lead dual in-line plastic package. It performs all the functions needed for the TV sound channel:

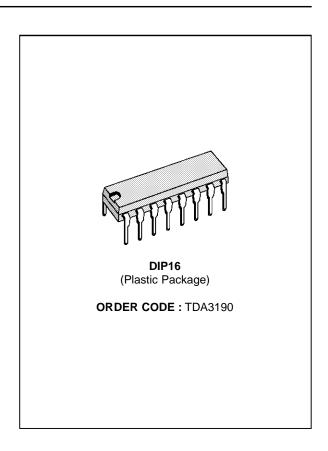
- IF LIMITER AMPLIFIER
- ACTIVE LOW-PASS FILTER
- FM DETECTOR
- DC VOLUME CONTROL
- AF PREAMPLIFIER
- AF OUTPUT STAGE

DESCRIPTION

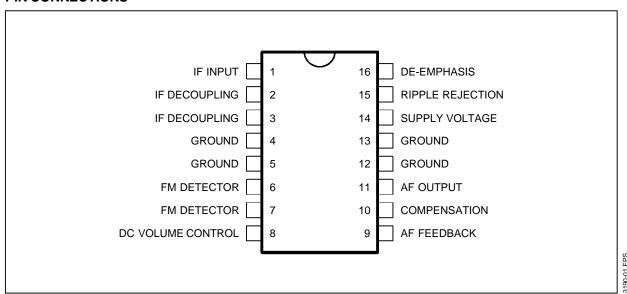
The TDA3190 can give an output power of 4.2 W (d = 10 %) into a 16 Ω load at V_S = 24 V, or 1.5 W (d = 10 %) into an 8 Ω load at V_S = 12 V. This performance, together with the FM-IF section characteristics of high sensitivity, high AM rejection and low distortion, enables the device to be used in almost every type of television receivers.

The device has no irradiation problems, hence no external screening is needed.

The TDA3190 is a pin to pin replacement of TDA1190Z.

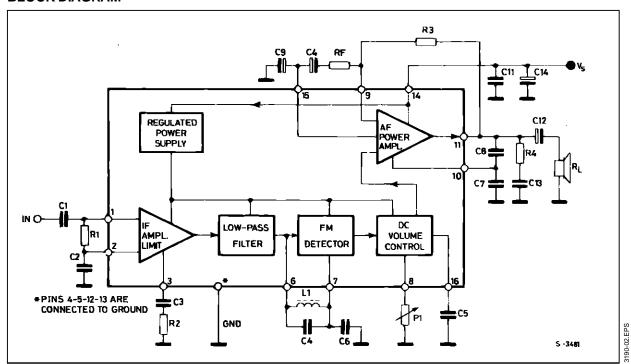


PIN CONNECTIONS



December 1992 1/9

BLOCK DIAGRAM



ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
Vs	Supply Voltage (pin 10)	28	V
Vi	Input Signal Voltage (pin 1)	1	V
Io	Output Peak Current (non-repetitive)	2	Α
Io	Output Peak Current (repetitive)	1.5	Α
P _{tot}	Power Dissipation at T _{pins} = 90 °C at T _{amb} = 70 °C (free air)	4.3 1	W
T _{stg} , T _j	Storage and Junction Temperature	- 40 to 150	°C

THERMAL DATA

Symbol	Parameter	Value	Unit] _
R _{th j-pins}	Thermal Resistance Junction-pins Max	14	°C/W	02.TE
R _{th j-amb}	Thermal Resistance Junction-ambient Max	80*	°C/W	3190

^{*} Obtained with the GND pins soldered to printed circuit with minimized copper area.

ELECTRICAL CHARACTERISTICS

(refer to the test circuit, $V_S = 24V$, $T_{amb} = 25$ °C unless otherwise specified)

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
Vs	Supply Voltage (Pin 14)		9		28	V
Vo	Quiescent Output Voltage (Pin11)	V _s = 24V V _s = 12V	11 5.1	12 6	13 6.9	< <
I _d	Quiescent Drain Current	$\begin{array}{l} P_1 = 22k\Omega \\ V_s = 24V \\ V_s = 12V \end{array}$	11	22 19	45 40	mA mA

1190-03.TBL

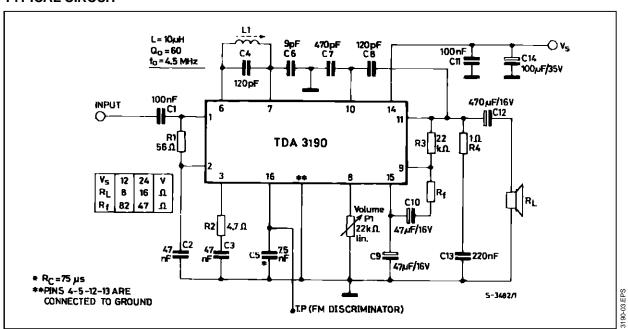


ELECTRICAL CHARACTERISTICS

(refer to the test circuit, $V_S = 24V$, $T_{amb} = 25^{\circ}C$ unless otherwise specified)

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
Po	Output Power	$\begin{array}{l} d = 10\%, f_m = 400 Hz, \\ f_0 = 4.5 MHz, \Delta f = \pm 25 kHz \\ V_s = 24 V, R_L = 16 \Omega \\ V_s = 12 V, R_L = 8 \Omega \end{array}$		4.2 1.5		W
		$\begin{array}{l} d=2\%,f_{m}=400Hz,\\ f_{o}=4.5MHz,\Delta f=\pm25kHz\\ V_{s}=24V,R_{L}=16\Omega\\ V_{s}=12V,R_{L}=8\Omega \end{array}$		3.5 1.4		W W
Vi	Input Limiting Voltage (–3dB) atPin 1	$f_0 = 4.5 MHz, \Delta f = \pm 7.5 kHz, f_m = 400 Hz, P_1 = 0$		40	100	μV
d	Distortion	$\begin{aligned} P_{o} &= 50 \text{mW}, f_{m} = 400 \text{Hz}, \\ f_{o} &= 4.5 \text{MHz}, \Delta f = \pm 7.5 \text{kHz} \\ V_{s} &= 24 \text{V}, R_{L} = 16 \Omega \\ V_{s} &= 12 \text{V}, R_{L} = 8 \Omega \end{aligned}$		0.75 1		% %
В	Frequency Response of audio amplifier (–3dB)	$ \begin{array}{l} R_L = 16\Omega, \ C_8 = 120 pF \\ C_7 = 470 pF, \ P_1 = 22 k\Omega \\ R_f = 82\Omega \\ R_f = 47\Omega \end{array} $		70 to 1200 70 to 7000		Hz Hz
Vo	Recovered Audio Voltage (Pin16)	$V_i \ge 1 \text{mV}, \ f_0 = 4.5 \text{MHz} \ f_m = 400 \text{Hz}, \ \Delta f = \pm 7.5 \text{kHz}, \ P_1 = 0$		120		mV
AMR	Ampliture Modulation Rejection	$V_i \ge 1 mV$, $f_0 = 4.5 MHz$, $f_m = 400 Hz$, $\Delta_f = \pm 25 kHz$, $m = 0.3$		55		dB
$\frac{S+N}{N}$	Signal to Noise Ratio	$\begin{aligned} V_I \geq 1 m V, \ V_o = 4 V, \ f_o = 4.5 MHz, \\ f_m = 400 Hz, \ \Delta f = \pm \ 25 kHz \end{aligned}$	50	65		dB
R ₃	External Feedback Resistance (betweenPins9and11)				25	kΩ
Ri	Input Resistance (Pin1)	V _i = 1mV, fo = 4.5MHz		30		kΩ
Ci	Input Capacitance (Pin1)			5		pF
SVR	Supply Voltage Rejection	$R_L = 16\Omega$, $f_{ripple} = 120Hz$, $P_1 = 22k\Omega$		46		dB
A_{v}	DC Volume Control Attenuation	$P_1 = 12k\Omega$		90		dB

TYPICAL CIRCUIT



3190-04.TBL

Figure 1: Relative Audio Output Voltage and Output

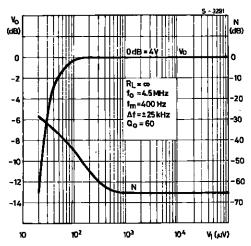


Figure 3: Amplitude Modulation Rejection versus Input Signal

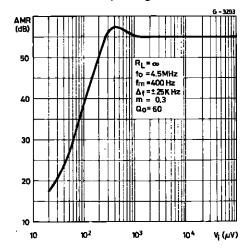


Figure 5: Recovered Audio Voltage versus Unloaded Q Factor of the Detector Coil

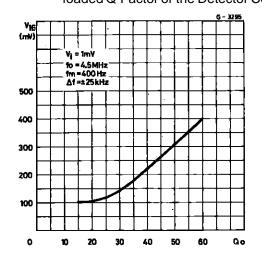


Figure 2 : Output Voltage Attenuation versus DC Volume Control Resistance

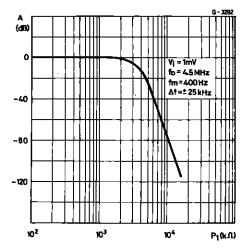


Figure 4 : ΔAMR versus Tuning Frequency Change

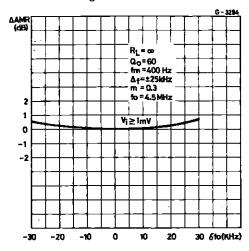
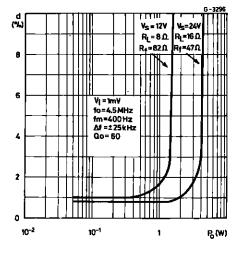


Figure 6: Distortion versus Output Power



90-05.EPS

190-07 FPS

90-09 FPS

3190-08.EPS

Figure 7: Distortion versus Frequency Deviation

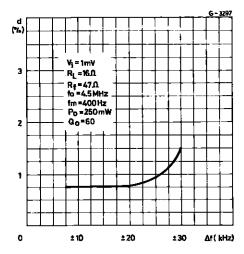


Figure 9: Audio Amplifier Frequency Response

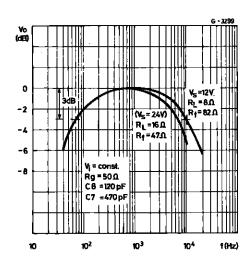


Figure 11: Supply Voltage Ripple Rejection versus Volume Control Attenuation

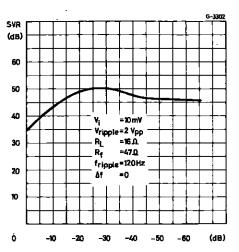


Figure 8: Distortion versus Tuning Frequency Change

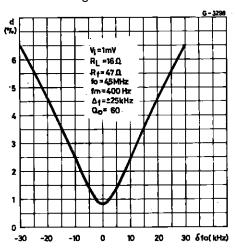


Figure 10: Supply Voltage Ripple Rejection versus Ripple Frequency

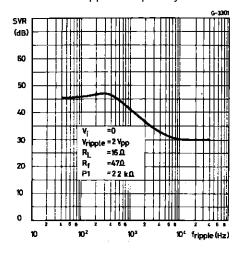


Figure 12: Output Power versus Supply Voltage

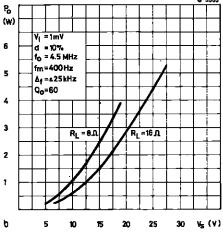


Figure 13: Maximum Power Dissipation versus Supply Voltage (sinewave operation)

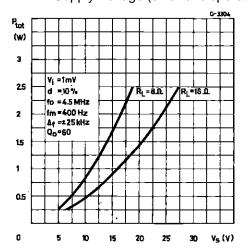


Figure 15: Quiescent Output Voltage (Pin 11) versus Supply Voltage

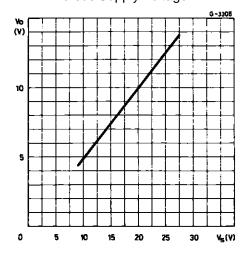
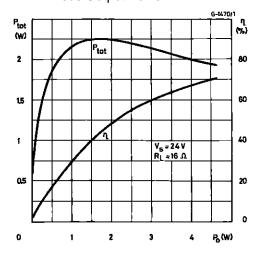


Figure 14: Power Dissipation and Efficiency versus Output Power



3190-17.EPS

10 10

APPLICATION INFORMATION

The electrical characteristics of the TDA3190 remain almost constant over the frequency range 4.5 to 6 MHz, therefore it can be used in all television standards (FM mod.). The TDA3190 has a high input impedance, so it can work with a ceramic filter or with a tuned circuit that provide the necessary input selectivity.

The value of the resistors connected to pin 9, determine the AC gain of the audio frequency amplifier. This enables the desired gain to be selected in relation to the frequency deviation at which the output stage of the AF amplifier, must enter into

clipping.

Capacitor C8, connected between pins 10 and 11, determines the upper cutoff frequency of the audio bandwidth. To increase the bandwidth the values of C8 and C7 must be reduced, keeping the ratio C7/C8 as shown in the table of fig. 16.

The capacitor connected between pin 16 and ground, together with the internal resistor of 10 K Ω forms the de-emphasis network. The Boucherot cell eliminates the high frequency oscillations caused by the inductive load and the wires connecting the loudspeaker.

Figure 16: Typical Application Circuit

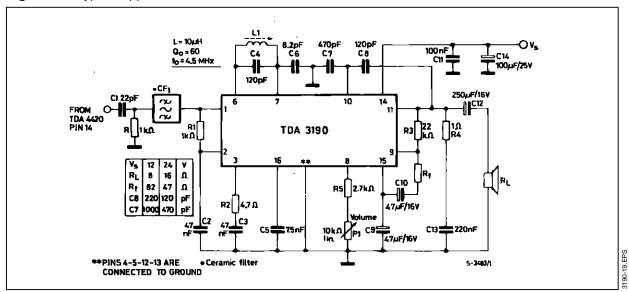
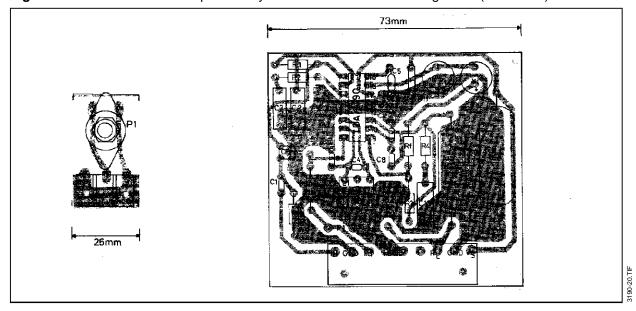


Figure 17: P.C. Board and Component Layout of the Circuit shown in Figure 16 (1:1 scale)



MOUNTING INSTRUCTION

The Rth j-amb of the TDA3190 can be reduced by soldering the GND pins to a suitable copper area of the printed circuit board (fig. 18) or to an external heatsink (fig. 19).

The diagram of figure 20 shows the maximum dissipable power Ptot and the Rth j-amb as a function of the side "I" of two equal square copper areas

Figure 18: Example of P.C. Board Copper Area which is used as Heatsink

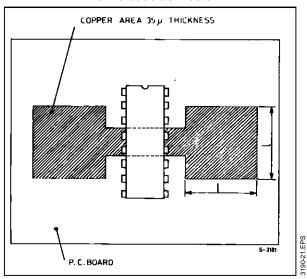
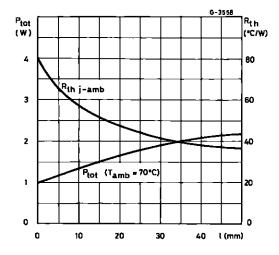


Figure 20: Maximum Dissipable Power and Junction to Ambient Thermal Resistance versus Side "T"



having a thickness of 35μ (1.4 mils).

During soldering the pins temperature must not exceed 260 °C and the soldering time must not be longer than 12 seconds.

The external heatsink or printed circuit copper area must be connected to electrical ground.

Figure 19: External Heatsink Mounting Example

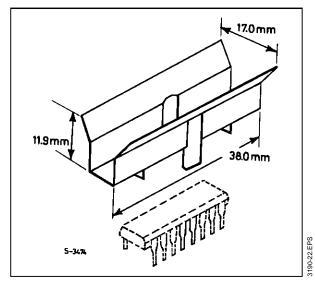
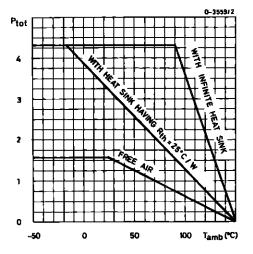


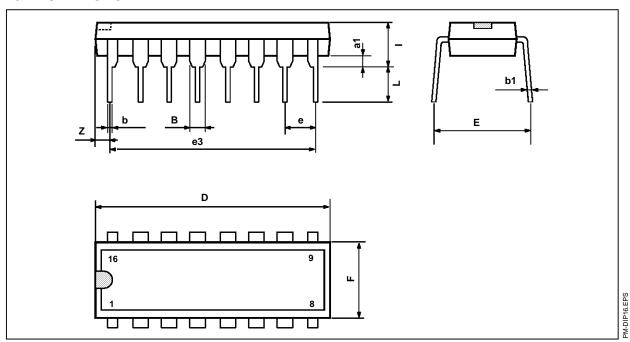
Figure 21: Maximum Allowable Power Dissipation versus Ambient Temperature



0-24.EPS

PACKAGE MECHANICAL DATA

16 PINS - PLASTIC DIP



Dimensions	Millimeters			Inches			
	Min.	Тур.	Max.	Min.	Тур.	Max.	
a1	0.51			0.020			
В	0.77		1.65	0.030		0.065	
b		0.5			0.020		
b1		0.25			0.010		
D			20			0.787	
E		8.5			0.335		
е		2.54			0.100		
e3		17.78			0.700		
F			7.1			0.280	
i			5.1			0.201	
L		3.3			0.130		
Z			1.27			0.050	

D16 TBI

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